

CSE311 Microwave Engineering

LEC (08)

Impedance Matching

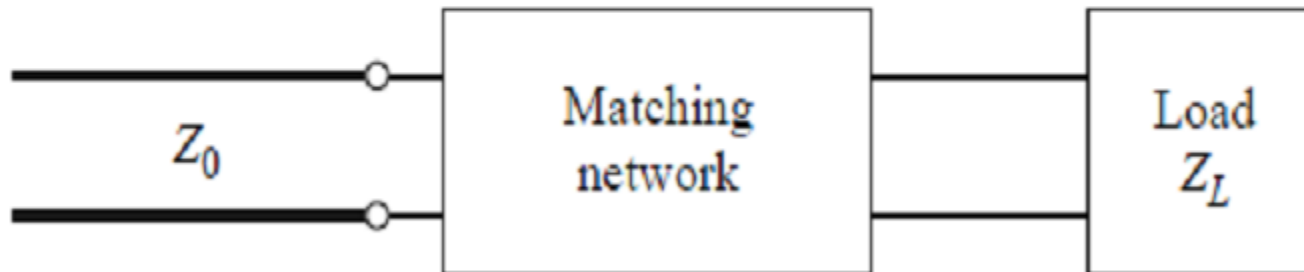
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3.7 Matching Networks

- The basic idea of impedance matching is illustrated in Figure 4.1, which shows an **impedance matching** network placed between a **load impedance** and a **transmission line**. The **matching network** is ideally **lossless**, to **avoid unnecessary loss of power**, and is usually designed so that the **impedance seen looking into the matching network is Z_0** .
- This procedure is sometimes referred to as **tuning**.



Impedance matching or tuning is important for the following reasons:

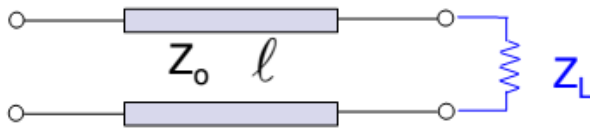
- **Maximum power is delivered** when the load is matched to the line (assuming the generator is matched), and **power loss in the feed line is minimized**.

3.7 Matching Networks

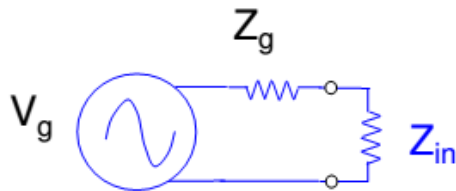
Why match?

$$\Gamma = \frac{Z - Z_o}{Z + Z_o}$$

Minimize reflection, maximize transmission

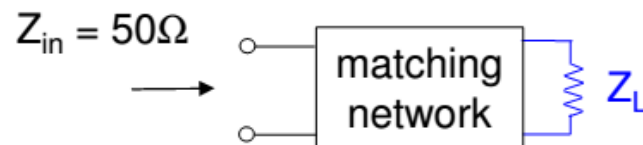


If $Z_L = Z_o$, line length is not important.



If $Z_{in} = Z_g$, max power to the load.

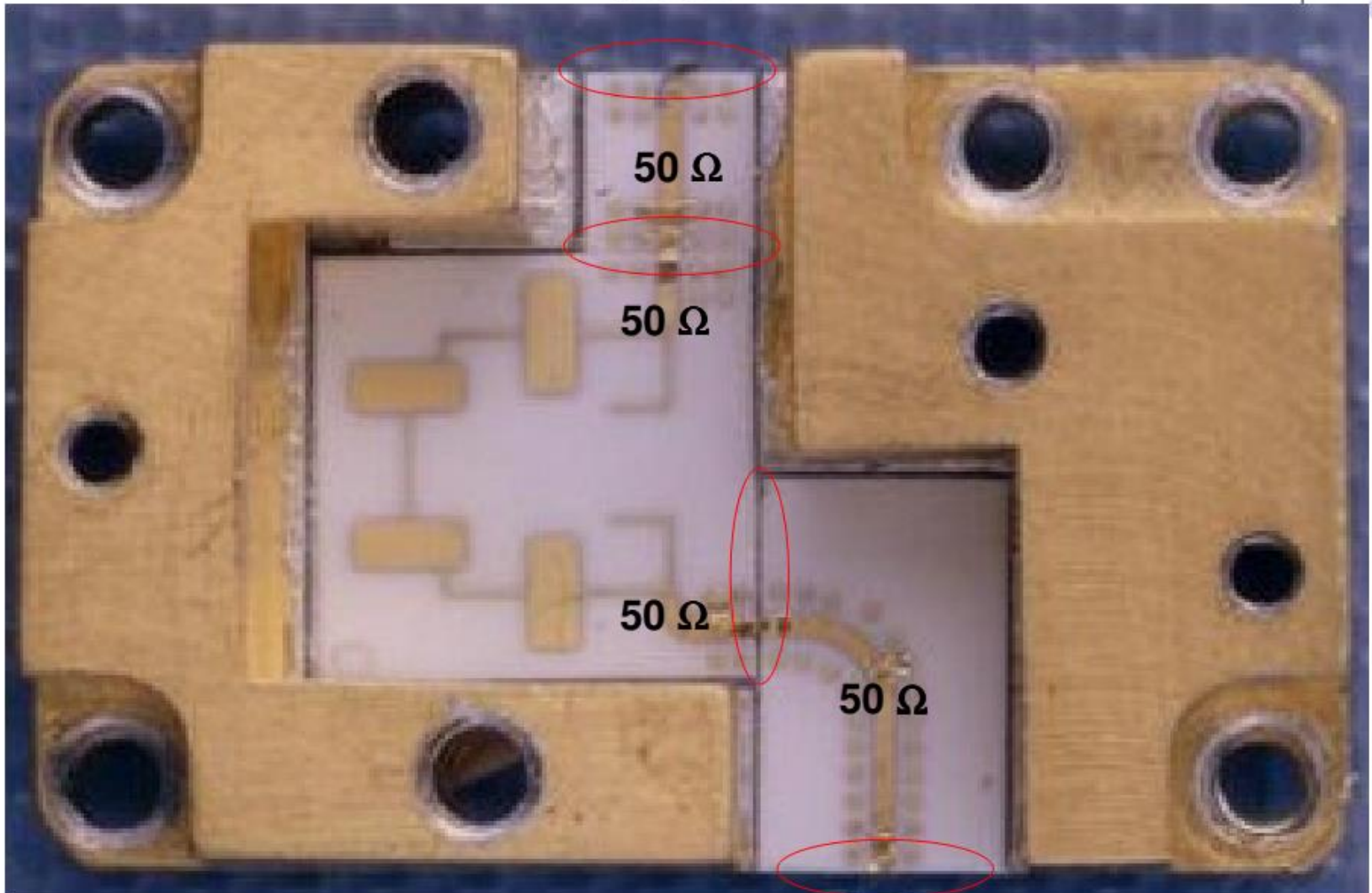
Goal: to design matching networks so all Z 's are the same = 50Ω



3.7 Matching Networks

Technician tuning even for 50Ω designs

input 50Ω

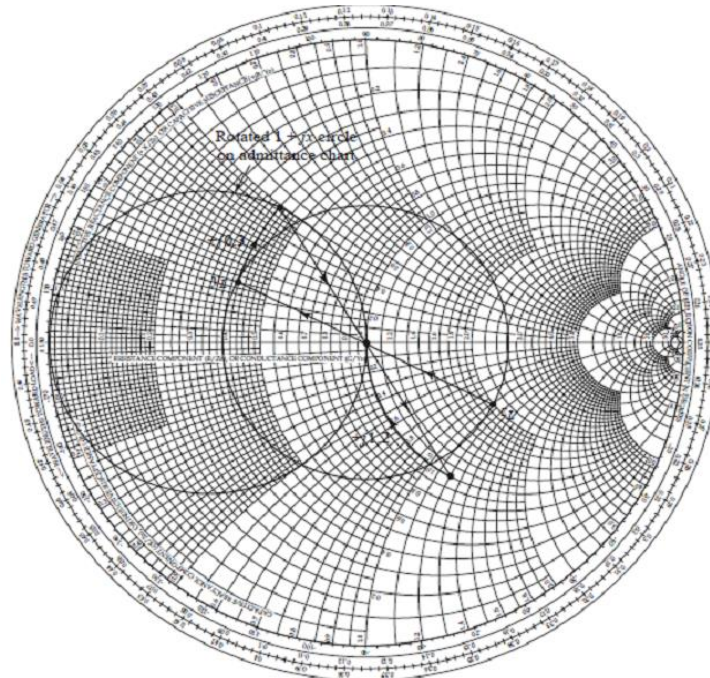
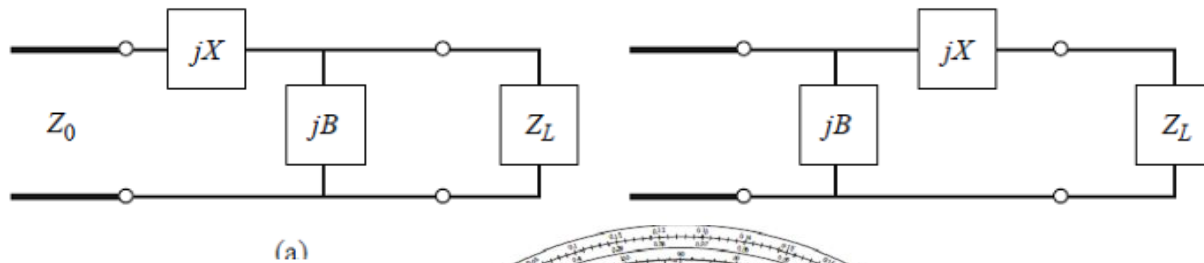


output 50Ω

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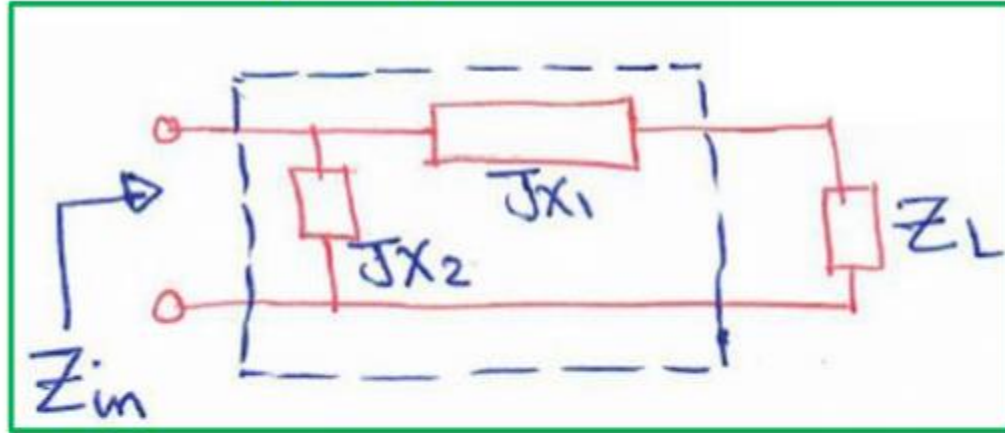
Matching with Lumped Elements (L – Networks)

- Probably the simplest type of **matching network is the L-section**, which uses **two reactive elements** to match an arbitrary load impedance to a transmission line.
- There are **two possible configurations** for this network, as shown in Figure 4.2.



3.7 Matching Networks

- Analytic Solution for $Z_L < Z_0$



$$Z_{in} = (Z_L + jX_1) \parallel jX_2 = Z_0$$

$$\frac{1}{Z_L + jX_1} + \frac{1}{jX_2} = \frac{1}{Z_0}$$

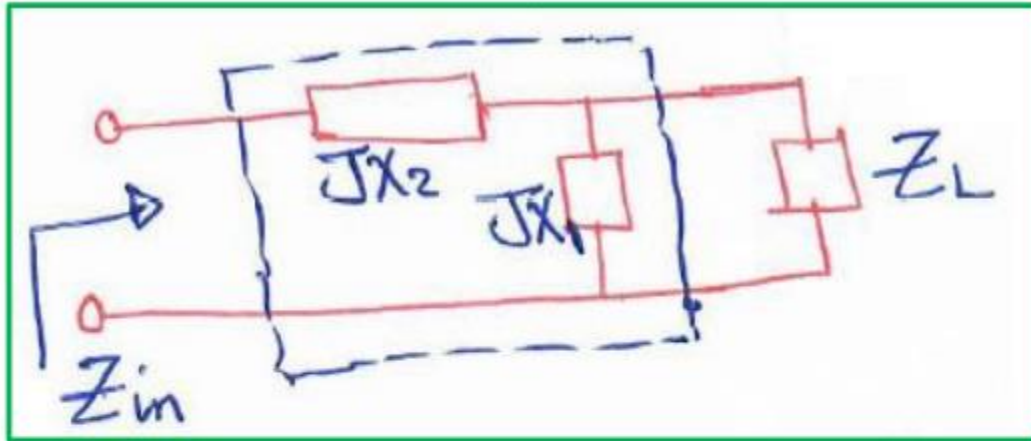
$$\frac{1}{Z_L + jX_1} = \boxed{\frac{1}{Z_0}} + \boxed{\frac{j}{X_2}}$$

real imaginary

- Rearranging and separating into real and imaginary parts gives:

3.7 Matching Networks

- Analytic Solution for $Z_L > Z_0$



- Rearranging and separating into real and imaginary parts gives:

$$Z_{in} = jX_2 + \frac{jX_1 * Z_L}{jX_1 + Z_L} = Z_0$$

$$\frac{jX_1 * Z_L}{jX_1 + Z_L} = \boxed{Z_0} - \boxed{jX_2}$$

real imaginary

3.7 Matching Networks

Example 1

Design an L-section matching Circuit for $Z_L = 60 + j40$ for $Z_0 = 100 \Omega$ and $f = 1\text{MHz}$.

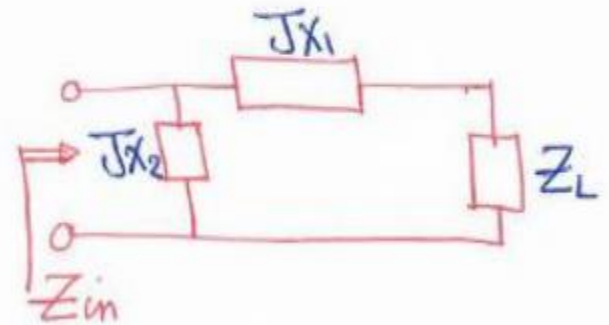
Solution

$$Z_L = 60 + j40 \Omega \quad \text{and} \quad Z_0 = 100 \Omega$$

$$\therefore \underline{\underline{Z_L < Z_0}}$$

$$\frac{1}{Z_L + jX_1} = \frac{1}{60 + j40 + jX_1}$$

$$\frac{1}{Z_L + jX_1} = \frac{1}{60 + j(40 + X_1)} \neq \frac{60 - j(40 + X_1)}{60 - j(40 + X_1)}$$



3.7 Matching Networks

Example 1

$$\frac{1}{Z_L + jX_1} = \frac{60 - j(40 + X_1)}{(60)^2 + (40 + X_1)^2}$$

For real part

$$\frac{60}{(60)^2 + (40 + X_1)^2} = \frac{1}{Z_0} = \frac{1}{100}$$

$$X_1^2 + 80X_1 + (60)^2 + (40)^2 - 60 \times 100 = 0$$

$$\boxed{X_1 = 8,99} \quad \text{or} \quad \boxed{X_1 = -88,99}$$

For imaginary part

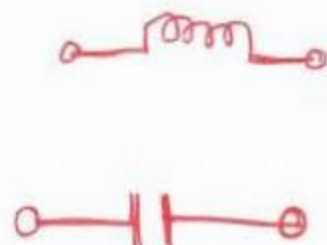
$$\frac{j}{X_2} = \frac{-j(40 + X_1)}{(60)^2 + (40 + X_1)^2}$$

$$\boxed{X_2 = -122,5} \quad \text{or} \quad \boxed{X_2 = 122,5}$$

3.7 Matching Networks

Example 1

N.B $JX = \begin{bmatrix} J\omega L \\ \frac{-J}{\omega C} \end{bmatrix}$



First Solution

$$X_1 = 8,99 \quad \text{and} \quad X_2 = -122,5$$

$$JX_1 = J\omega L$$

$$L = \frac{8,99}{2\pi f} = \frac{8,99}{2\pi \times 10^6}$$

$$L = 1,43 \mu\text{H}$$

$$JX_2 = \frac{-J}{\omega C}$$

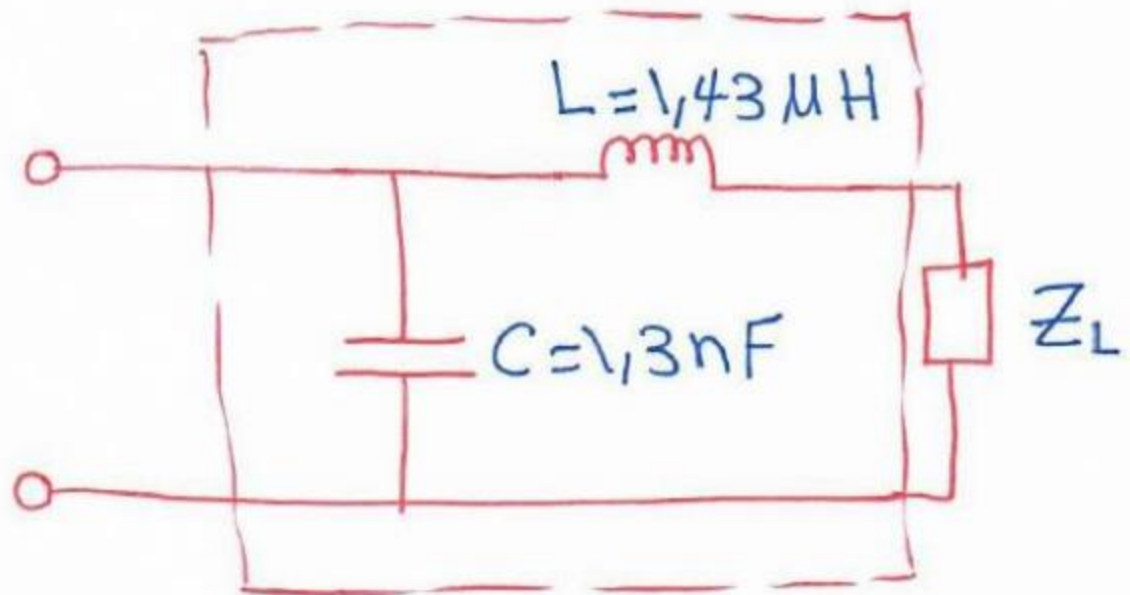
$$122,5 = \frac{1}{2\pi \times 10^6 \times C}$$

$$C = 1,3 \text{ nF}$$

3.7 Matching Networks

Example 1

matching Circuit



3.7 Matching Networks

Example 1

Second Solution

$$X_1 = -88,99$$

$$jX_1 = \frac{-j}{\omega C}$$

$$88,99 = \frac{1}{2\pi \times 10^6 \times C}$$

$$C = 1,79 \text{ nF}$$

and

$$X_2 = 122,5$$

$$jX_2 = j\omega L$$

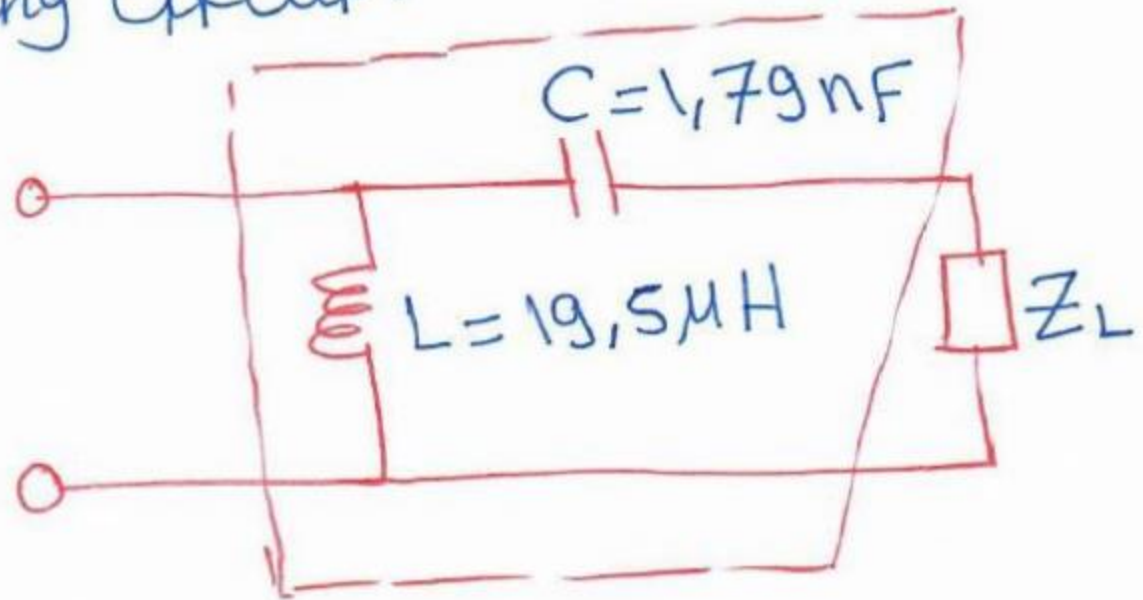
$$L = \frac{122,5}{2\pi \times 10^6}$$

$$L = 19,5 \mu\text{H}$$

3.7 Matching Networks

Example 1

matching Circuit



3.7 Matching Networks

Example 2 Design an L-Section matching Circuit for $Z_L = 150 + j40$ with $Z_0 = 100\Omega$ and $f = 1\text{MHz}$

Solution

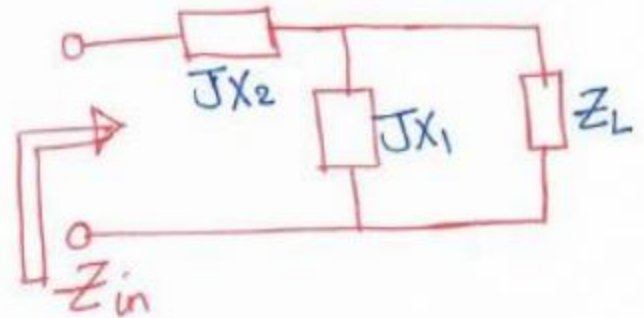
$$Z_L = 150 + j40\Omega \text{ and } Z_0 = 100\Omega$$

$$\therefore \underline{\underline{Z_L > Z_0}}$$

$$Z_{in} = jX_2 + \frac{jX_1 * Z_L}{Z_L + jX_1}$$

$$Z_{in} = Z_0$$

$$\frac{jX_1 * Z_L}{jX_1 + Z_L} = Z_0 - jX_2$$



3.7 Matching Networks

Example 2

$$\begin{aligned}\therefore \frac{jX_1(150 + j40)}{150 + j40 + jX_1} &\neq \frac{150 - j(40 + X_1)}{150 - (40 + X_1)} = \\ &= \frac{jX_1[(150)^2 + j6000 - j150(40 + X_1) + 40(40 + X_1)]}{(150)^2 + (40 + X_1)^2} \\ &= \frac{150 X_1^2 + j[X_1 \neq (150)^2 + 1600X_1 + 40X_1^2]}{(150)^2 + (40 + X_1)^2}\end{aligned}$$

3.7 Matching Networks

Example 2

For real part

$$\frac{150 X_1^2}{(150)^2 + (40 + X_1)^2} = Z_0 = 100$$

$$100 [(150)^2 + (40 + X_1)^2] = 150 X_1^2$$

$$2 [(150)^2 + (40)^2 + 80 X_1 + X_1^2] = 3 X_1^2$$

$$X_1^2 - 160 X_1 - 2[(150)^2 + (40)^2] = 0$$

$$X_1 = 313,67$$

or

$$X_1 = -153,67$$

For imaginary part

$$-jX_2 = \frac{(150)^2 X_1 + 1600 X_1 + 40 X_1^2}{(150)^2 + (40 + X_1)^2}$$

$$X_2 = 77,89$$

or

$$X_2 = -77,89$$

3.7 Matching Networks

Example 2

First Solution

$$X_1 = 313,67$$

and

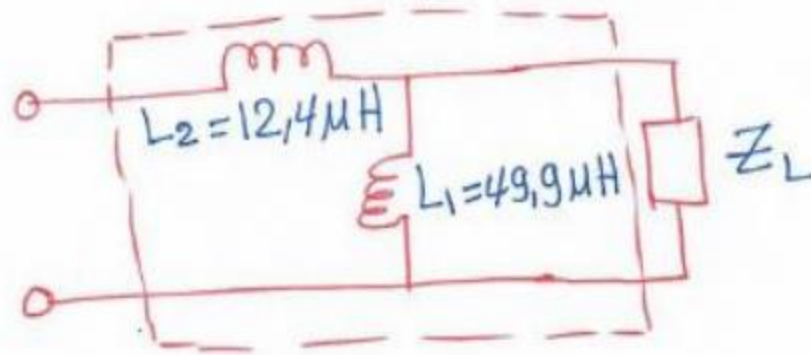
$$X_2 = 77,89$$

$$L_1 = \frac{313,67}{2\pi \times f}$$

$$L_1 = 49,9 \mu\text{H}$$

$$L_2 = \frac{77,89}{2\pi \times f}$$

$$L_2 = 12,4 \mu\text{H}$$



3.7 Matching Networks

Example 2

Second Solution

$$X_1 = -153,67$$

and

$$X_2 = -77,89$$

$$C_1 = \frac{1}{153,67 \times 2\pi \times f}$$

$$C_2 = \frac{1}{77,89 \times 2\pi \times f}$$

$$C_1 = 1,036 \text{ nF}$$

$$C_2 = 2,044 \text{ nF}$$

